

# MATLAB Function to Design Second Order Filters

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## 1 Introduction

Second order filters, sometimes called *second order sections* (SOS), are useful building blocks for audio signal processing. Three common SOS types are

- Bass shelf filter - boosts or cuts frequency response around and below a transition frequency.
- Treble shelf filter - boosts or cuts frequency response around and above a transition frequency.
- Peak filter - boosts or cuts frequency response around a peak frequency.

At other frequencies, e.g., above the transition frequency for a bass shelf filter, an SOS is transparent, i.e., the amplitude of the frequency response is 0 dB. Thus, a cascade of SOS can boost or cut frequency response at several frequencies. This property makes a SOS an attractive component for an audio equalizer.

The MATLAB function `order2filt` uses a straightforward algorithm to determine the coefficients of a SOS based on the filter type, sampling frequency, dB gain, peak or transition frequency, and a Q factor.

## 2 Filter Design Algorithm

We can use a two-step process to design a digital filter:

1. Obtain the transfer function for the corresponding analog filter  $H(s)$ .
2. Apply the bilinear transform to determine the digital filter transfer function  $H(z)$ .

The MATLAB function `bilinear`, which is included in the MATLAB signal processing toolbox, accomplishes the second step, so we need only determine  $H(s)$ .

## 2.1 Shelf Filters

Consider a biquadratic transfer function with poles and zeros given by

$$\begin{aligned} p &= -\sqrt{b}e^{\pm j\pi/4} \\ z &= -\sqrt{a}e^{\pm j\pi/4} \end{aligned} \quad (1)$$

The transfer function is given by

$$\begin{aligned} H(s) &= K \frac{(s + \sqrt{a}e^{j\pi/4})(s + \sqrt{a}e^{-j\pi/4})}{(s + \sqrt{b}e^{j\pi/4})(s + \sqrt{b}e^{-j\pi/4})} \\ &= K \frac{s^2 + \sqrt{2a}s + a}{s^2 + \sqrt{2b}s + b} \end{aligned} \quad (2)$$

where  $K$  is an arbitrary constant. If we choose  $K = b$ , then the magnitude of  $H(s)$  is  $a$  at low frequencies and  $b$  at high frequencies. For  $a = 1$  and  $b = g$ ,  $H(s)$  is a treble shelf filter with gain  $g$ . For  $a = g$  and  $b = 1$ ,  $H(s)$  is a bass shelf filter. Figure 1 shows the frequency response for  $a = 1$  and  $b = g$  (treble shelf filter) for several values of  $g$  (derived from given dB values). Note the lack of symmetry in the frequency response curves, e.g., the curve for  $-5$  dB is not the mirror image of the curve for  $+5$  dB. We can fix this by inverting  $H(s)$  and replacing  $g$  with  $1/g$  for  $g < 1$  ( $< 0$  dB). This is equivalent to calculating the gain factor  $g$  as

$$g = 10^{|G|/20} \quad (3)$$

and inverting  $H(s)$  for  $G < 0$ . The result is shown in Figure 2. Figure 3 shows a similar result for a bass shelf filter.

The poles and zeros of our shelf filters lie along a line at an angle of  $\pm\pi/4$  relative to the negative real axis. If we change this angle to a general angle  $\theta$ , we have

$$\begin{aligned} p &= -\sqrt{b}e^{\pm j\theta} \\ z &= -\sqrt{a}e^{\pm j\theta} \end{aligned} \quad (4)$$

and our transfer function becomes

$$H(s) = b \frac{s^2 + 2\sqrt{a}\cos\theta s + a}{s^2 + 2\sqrt{b}\cos\theta s + b} \quad (5)$$

Figure 4 shows the frequency response of a bass boost filter with gain  $G = 10$  dB and various values of  $\theta$ . As  $\theta$  approaches  $\pi/2$ , we get peaking, which is probably not desirable, and for  $\theta \geq \pi/2$ , the filter becomes unstable. However, for  $\theta$  up to about  $\pi/3$ , we get an increasing slope with no significant peaking.

Since the value of  $\cos\theta$  varies from 1 to 0 as  $\theta$  varies from 0 to  $\pi/2$ , we can also express  $H(s)$  as

$$H(s) = b \frac{s^2 + 2\sqrt{a}s/Q + a}{s^2 + 2\sqrt{b}s/Q + b} \quad (6)$$

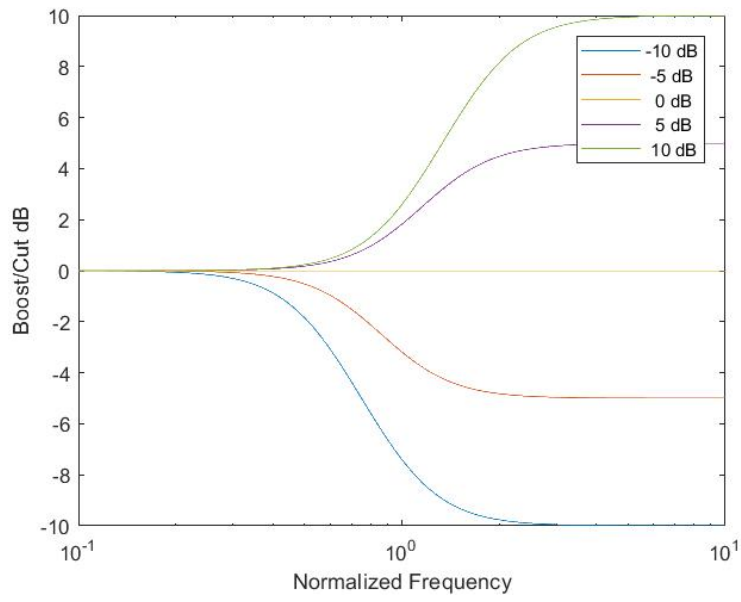


Figure 1: Frequency Response of Treble Shelf Filter

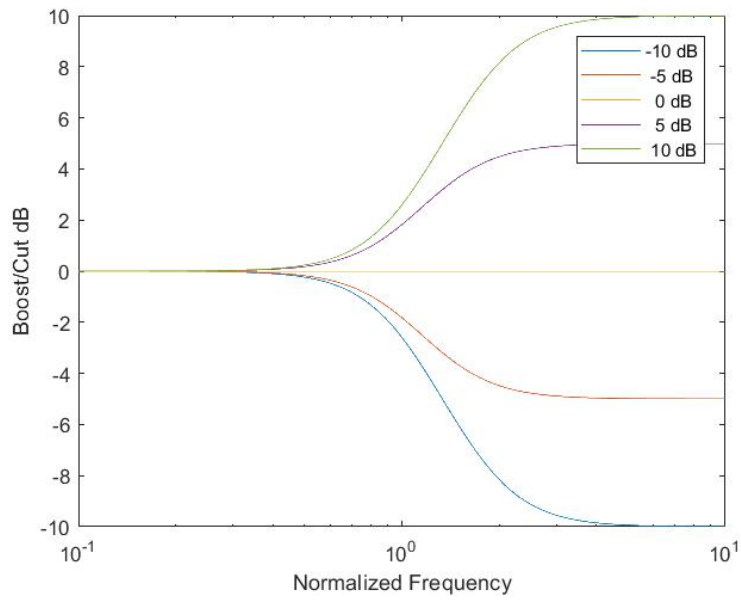


Figure 2: Frequency Response of Symmetric Treble Shelf Filter

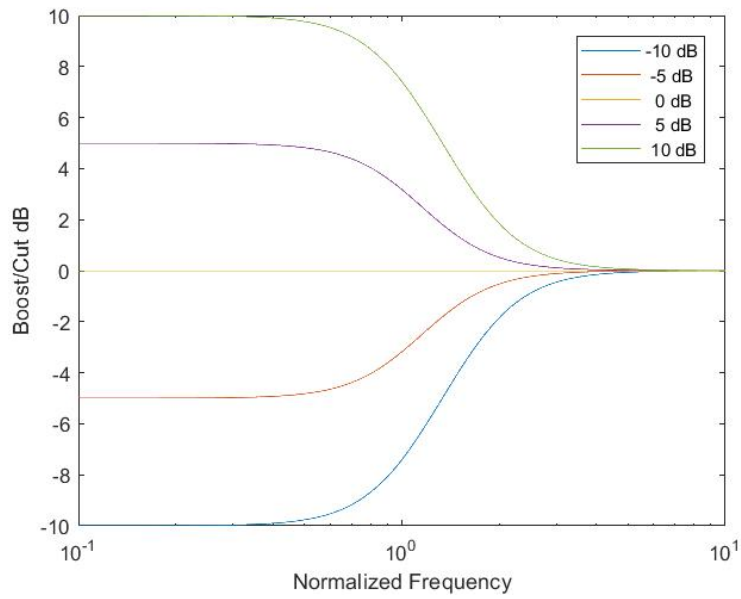


Figure 3: Frequency Response of Symmetric Bass Shelf Filter

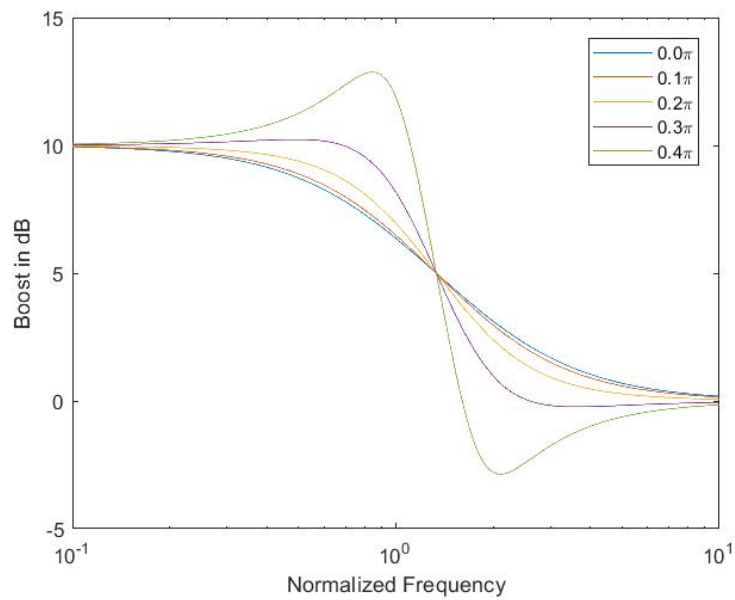


Figure 4: Frequency Response of Bass Boost Filter with Angle Parameter

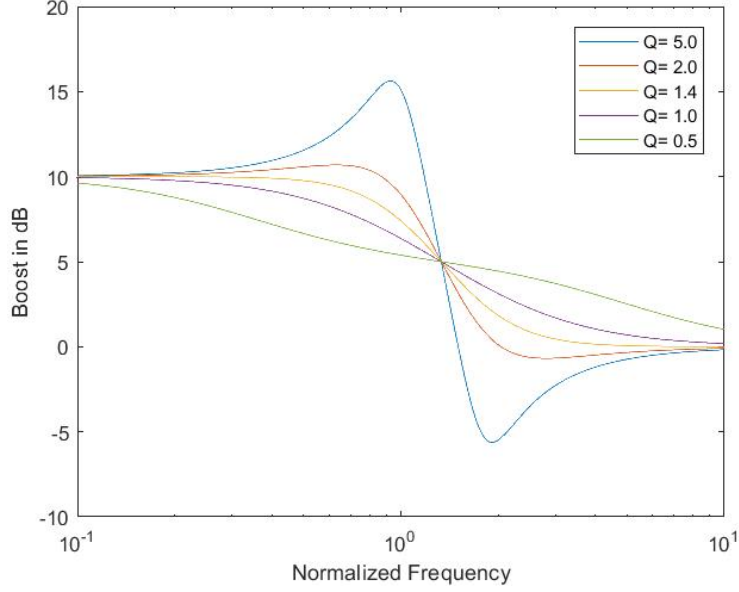


Figure 5: Frequency Response of Bass Boost Filter with Q Factor

where  $Q \geq 1$ . Peaking occurs for  $Q > \sqrt{2}$  and becomes significant for  $Q > 2$ . Figure 5 shows how the shelf filter frequency response varies with  $Q$ . Note that for  $Q < 1$ , the poles and zeros are on the negative real axis and the filter behaves more like a cascade of two first order filters.

The function `order2filt` derives shelf filters based on Equation 6, adjusting the transition frequency by replacing  $s$  with  $s/\omega_0$ . The shelf filter analog transfer functions are summarized here:

**Treble boost**

$$H(s) = g \frac{s^2 + 2\omega_0 s/Q + \omega_0^2}{s^2 + 2\sqrt{g}\omega_0 s/Q + g\omega_0^2} \quad (7)$$

**Treble cut**

$$H(s) = \frac{1}{g} \frac{s^2 + 2\sqrt{g}\omega_0 s/Q + g\omega_0^2}{s^2 + 2\omega_0 s/Q + \omega_0^2} \quad (8)$$

**Bass boost**

$$H(s) = \frac{s^2 + 2\sqrt{g}\omega_0 s/Q + g\omega_0^2}{s^2 + 2\omega_0 s/Q + \omega_0^2} \quad (9)$$

**Bass cut**

$$H(s) = \frac{s^2 + 2\omega_0 s/Q + \omega_0^2}{s^2 + 2\sqrt{g}\omega_0 s/Q + g\omega_0^2} \quad (10)$$

where  $\omega_0 = 2\pi f_0$ ,  $f_0$  is the transition frequency, and  $g = 10^{|G|/20}$ .

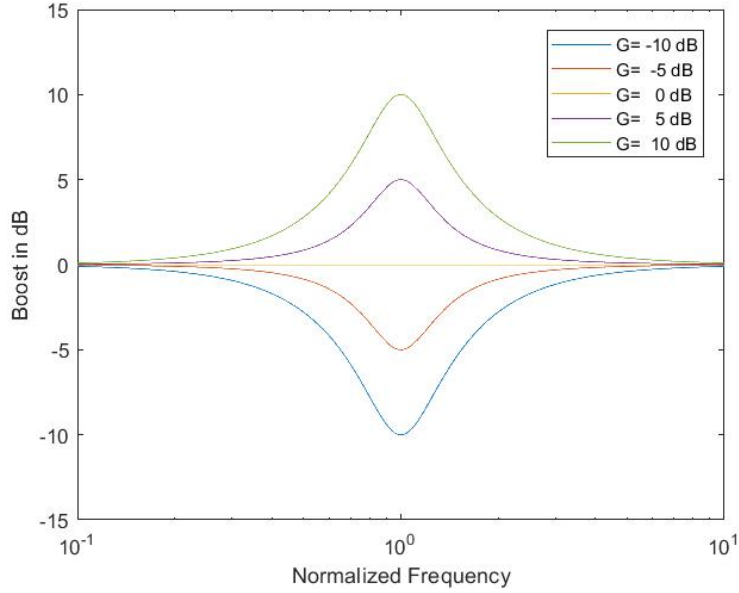


Figure 6: Frequency Response of Peak Filter with  $Q = 2$  and Varying Gain

## 2.2 Peak Filters

Now consider a biquadratic transfer function with poles and zeros given by

$$\begin{aligned} p &= -b \pm j\sqrt{1 - b^2} \\ z &= -a \pm j\sqrt{1 - a^2} \end{aligned} \quad (11)$$

For  $0 \leq a, b \leq 1$ , these poles and zeros lie along a circle of radius 1 in the left-hand complex plane. The transfer function is given by

$$H(s) = \frac{s^2 + 2as + 1}{s^2 + 2bs + 1} \quad (12)$$

where the constant factor has been set to  $K = 1$ . For  $a > b$ , the corresponding frequency response magnitude,  $|H(j\omega)|$ , reaches a peak value of  $a/b$  at normalized frequency  $\omega = 1$ . We will let  $a = g/2Q$  and  $b = 1/2Q$  so that

$$H(s) = \frac{s^2 + gs/Q + 1}{s^2 + s/Q + 1} \quad (13)$$

For  $g > 1$ ,  $H(s)$  is a peak filter with a maximum gain of  $g$  at normalized frequency  $\omega = 1$ . Again, if we calculate  $g$  from the absolute value of the dB gain  $G$  and invert  $H(s)$  for  $G < 0$ , we have a symmetric peak filter that can either boost or cut the signal. Figures 6 and 7 show the effects of both  $G$  and  $Q$  on the peak filter frequency response.

The analog transfer functions for a peak filter with peak frequency  $\omega_0$  become

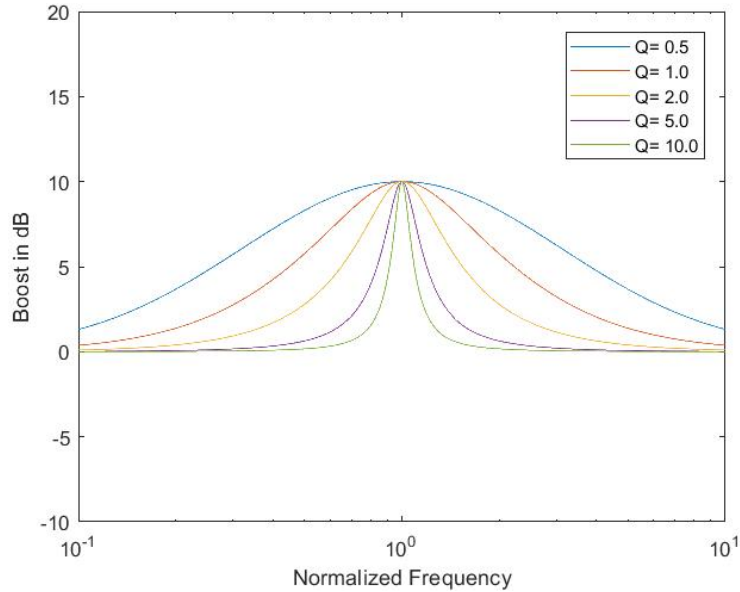


Figure 7: Frequency Response of Peak Filter with  $G = 10$  dB and Varying  $Q$

#### Peak Boost

$$H(s) = \frac{s^2 + g\omega_0 s/Q + \omega_0^2}{s^2 + \omega_0 s/Q + \omega_0^2} \quad (14)$$

#### Peak Cut

$$H(s) = \frac{s^2 + \omega_0 s/Q + \omega_0^2}{s^2 + g\omega_0 s/Q + \omega_0^2} \quad (15)$$

### 2.3 Bilinear Transform

As stated earlier, `order2filt` uses the MATLAB toolbox function `bilinear` to convert the analog filter coefficients to digital filter coefficients. The syntax is

`[b, a]= bilinear (B, A, fs , f0 ) ;`

where `b, a` are the digital filter coefficients, `B, A` are the analog filter coefficients, `fs` is the sampling rate, and `f0` is the peak or transition frequency. Called this way, `bilinear` uses warping to match the frequency responses at the peak or transition frequency. The analog filter coefficients are calculated from the inputs to `order2filter` (see next section) according to the transfer function formulas listed above. We need only use the boost formulas, since we can simply reverse `A` and `B` in the call to `bilinear` for  $G < 0$ .

## 3 The order2filt Function

### 3.1 Syntax

```
[b,a] = order2filt(ftype,fs,G,f0,Q);
```

### 3.2 Description

The function `order2filt` returns the `b` and `a` coefficients of a second order filter. The input `ftype` specifies the filter type, `fs` is the sampling frequency, `G` is the gain in dB, `f0` is the center or transition frequency, and `Q` is the  $Q$  factor.

### 3.3 Inputs

- `ftype` - filter type - 'Bass' or 'bass' for bass shelf filter, 'Treble' or 'treble' for treble shelf filter, or 'Peak' or 'peak' for peak filter
- `fs` - sampling rate in Hz (samples/second)
- `G` - gain in dB. For a peak filter, the gain at the peak frequency, for a bass shelf filter, the gain at frequencies sufficiently below the transition frequency, and for a treble shelf filter, the gain at frequencies sufficiently above the transition frequency
- `f0` - the peak frequency of a peak filter or the transition frequency of a shelf filter
- `Q` - the  $Q$  factor - adjusts the slope of a shelf filter and the width of a peak filter

### 3.4 Outputs

- `[b,a]` - the filter numerator and denominator coefficients - both `b` and `a` are 3-element vectors

### 3.5 Example

In this example, we use `order2filt` to create a four band equalizer (with all settings fixed) and display the magnitude of the frequency response. The MATLAB script is

```
fs = 44100;
ftype = {'Bass' 'Peak' 'Peak' 'Treble'};
Gain = [-2 1 0 2];
freq = [50 500 2000 5000];
Q = [1.2 1.5 1.5 sqrt(2)];
N = length(Gain);
sos = zeros(N,6);
for n=1:N
    [b,a] = order2filt(char(ftype(n)),fs,Gain(n),freq(n),Q(n));
    sos(n,:) = [b a];
end
[h,f] = freqz(sos,4096,fs);
semilogx(f,20*log10(abs(h)))
xlabel('Frequency_in_Hz')
```



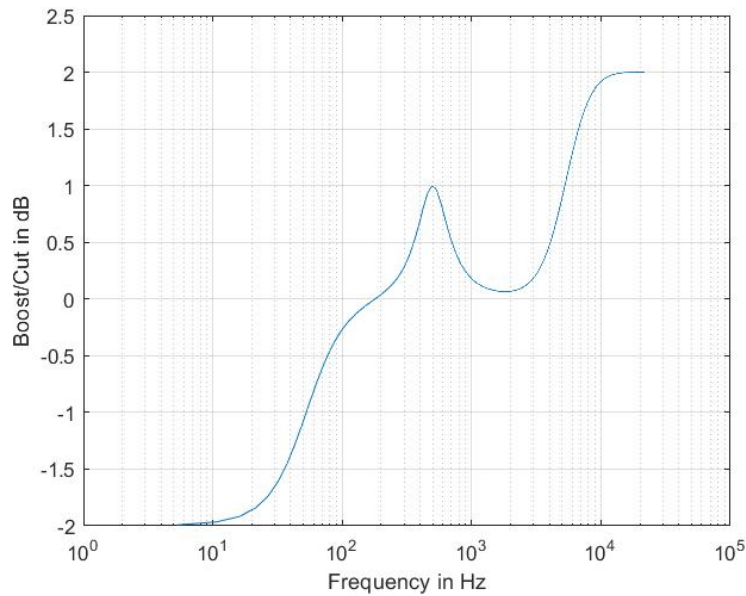


Figure 8: Frequency Response of Four Band Equalizer

```
ylabel( 'Boost/Cut in dB' )  
grid on
```

The equalizer is a cascade of four SOS (bass shelf, two peaks, and treble shelf) with dB gains, frequencies, and  $Q$  values specified. The four sets of coefficients are calculated using `order2filt` and loaded into an SOS matrix, which organizes the coefficients into four rows of six coefficients. The MATLAB function `freqz` accepts an SOS matrix as a possible input. Figure 8 shows the result.

With the MATLAB DSP system toolbox, you can also use the SOS matrix to create a `dsp.BiquadFilter` object, which can filter a stream of audio sample blocks.